# A Frequency-Domain Condition for the Stability of Feedback Systems Containing a Single Time-Varying Nonlinear Element

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It is proved that a condition similar to the Nyquist criterion guarantees the stability (in an important sense) of a large class of feedback systems containing a single time-varying nonlinear element. In the case of principal interest, the condition is satisfied if the locus of a certain complex-valued function (a) is bounded away from a particular disk located in the complex plane, and (b) does not encircle the disk.

## I. INTRODUCTION

The now well-known techniques introduced by Lyapunov have led to many very interesting results concerning the stability of time-varying nonlinear feedback systems governed by systems of differential equations. However, these methods have by no means led to a definitive theory of stability for even the simplest nontrivial time-varying nonlinear feedback systems. The general problem is, of course, one of considerable difficulty.

The unparalleled utility of the Nyquist stability criterion for single-loop, linear, time-invariant feedback systems is directly attributable to the fact that it is an explicit frequency-domain condition. The Nyquist locus not only indicates the stability or instability of a system, it presents the information in such a way as to aid the designer in arriving at a suitable design. The criterion is useful even in cases in which the system is so complicated that a sufficiently accurate analysis is not feasible, since experimental measurements can be used to construct the loop-gain locus.

The primary purpose of this article is to point out that some recently

obtained mathematical results,¹ not involving the theory of Lyapunov, imply that a condition similar to, and possessing the advantages of, the Nyquist criterion guarantees the stability (in an important sense) of feedback systems containing a single time-varying nonlinear element.\*•†

## II. THE PHYSICAL SYSTEM AND DEFINITION OF $\mathfrak{L}_2$ -STABILITY

Consider the feedback system of Fig. 1. We shall restrict our discussion throughout to cases in which  $g_1$ , f, u, and v denote real-valued measurable functions of t defined for  $t \ge 0$ .

The block labeled  $\psi$  is assumed to represent a memoryless timevarying (not necessarily linear) element that introduces the constraint  $u(t) = \psi[f(t),t]$ , in which  $\psi(x,t)$  is a function of x and t with the

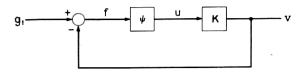


Fig. 1 — Nonlinear feedback system.

properties that  $\psi(0,t) = 0$  for  $t \ge 0$  and there exist a positive constant  $\beta$  and a real constant  $\alpha$  such that

$$\alpha \le \frac{\psi(x,t)}{x} \le \beta, \qquad t \ge 0$$

for all real  $x \neq 0$ . In particular, we permit the extreme cases in which  $\psi(x,t)$  is either independent of t or linear in x [i.e.,  $\psi(x,t) = \psi(1,t)x$ ].

The block labeled K represents the linear time-invariant portion of the forward path. It is assumed to introduce the constraint

$$v(t) = \int_0^t k(t-\tau)u(\tau)d\tau - g_2(t), \qquad t \ge 0$$

in which k and  $g_2$  are real-valued functions such that

$$\int_0^\infty |k(t)| dt < \infty, \qquad \int_0^\infty |g_2(t)|^2 dt < \infty. \tag{1}$$

† For results concerned with frequency-domain conditions for the global asymptotic stability (a sense of stability that is different from the one considered here) of nonlinear systems, see, for example, Refs. 2–4.

<sup>\*</sup> The results of Ref. 1 relate to feedback systems containing an arbitrary finite number of time-varying nonlinear elements, but, with the exception of the case discussed here, they do not admit of a simple geometric interpretation.

† For results concerned with frequency-domain conditions for the global asymp-

The function  $g_2$  takes into account the initial conditions at t = 0. Our assumptions regarding **K** are satisfied, for example, if, as is often the case, u and v are related by a differential equation of the form

$$\sum_{n=0}^{N} a_n \frac{d^n v}{dt^n} = \sum_{n=0}^{N-1} b_n \frac{d^n u}{dt^n}, \qquad t \ge 0$$

in which the  $a_n$  and the  $b_n$  are constants with  $a_N \neq 0$ , and

$$\sum_{n=0}^{N} a_n s^n \neq 0 \quad \text{for} \quad \text{Re}[s] \ge 0.$$

However, we do not require that u and v be related by a differential equation (or by a system of differential equations).

Assumption: We shall assume throughout that the response v is well defined and satisfies the inequality

$$\int_0^t |v(\tau)|^2 d\tau < \infty \tag{2}$$

for all finite t > 0, for each initial-condition function  $g_2$  that meets the conditions stated above and each input  $g_1$  such that

$$\int_0^\infty |g_1(t)|^2 dt < \infty.$$

Although this assumption plays an important role in the proof of the theorem to be presented, from an engineering viewpoint it is a trivial restriction (see Ref. 5).

Definition: We shall say that the feedback system of Fig. 1 is " $\mathfrak{L}_2$ -stable" if and only if there exists a positive constant  $\rho$  with the property that the response v satisfies

$$\left(\int_{0}^{\infty} |v(t)|^{2} dt\right)^{\frac{1}{2}} \leq \rho \left(\int_{0}^{\infty} |g_{1}(t)| + |g_{2}(t)|^{2} dt\right)^{\frac{1}{2}} + \left(\int_{0}^{\infty} |g_{2}(t)|^{2} dt\right)^{\frac{1}{2}}$$

for every initial-condition function  $g_2$  that meets the conditions stated above, and every input  $g_1$  such that

$$\int_0^\infty |g_1(t)|^2 dt < \infty.$$

In particular, if the system is  $\mathcal{L}_2$ -stable, then the response is square-integrable whenever the input is square-integrable.

It can be shown\* that the response v(t) approaches zero as  $t \to \infty$  for any square-integrable input  $g_1$ , provided that the system is  $\mathcal{L}_2$ -stable,

<sup>\*</sup> See the proof of Theorem 6 of Ref. 1.

 $g_2(t) \to 0$  as  $t \to \infty$ , and

$$\int_0^\infty |k(t)|^2 dt < \infty. \tag{3}$$

In addition, it follows at once from the Schwarz inequality that the response v(t) is uniformly bounded on  $[0, \infty)$  for any square-integrable input  $g_1$ , provided that the system is  $\mathfrak{L}_2$ -stable,  $g_2(t)$  is uniformly bounded on  $[0, \infty)$ , and (3) is satisfied.

# III. SUFFICIENT CONDITIONS FOR \$\mathcal{L}\_2\$-STABILITY

Theorem: Let

$$K(i\omega) = \int_0^\infty k(t)e^{-i\omega t}dt, \quad -\infty < \omega < \infty.$$

The feedback system of Fig. 1 is  $\mathfrak{L}_2$ -stable if one of the following three conditions is satisfied:

(i)  $\alpha > 0$ ; and the locus of  $K(i\omega)$  for  $-\infty < \omega < \infty$  (a) lies outside the circle  $C_1$  of radius  $\frac{1}{2}(\alpha^{-1} - \beta^{-1})$  centered on the real axis of the complex plane at  $[-\frac{1}{2}(\alpha^{-1} + \beta^{-1}), 0]$ , and (b) does not encircle  $C_1$  (see Fig. 2)

(ii)  $\alpha = 0$ , and  $Re[K(i\omega)] > -\beta^{-1}$  for all real  $\omega$ 

(iii)  $\alpha < 0$ , and the locus of  $K(i\omega)$  for  $-\infty < \omega < \infty$  is contained within the circle  $C_2$  of radius  $\frac{1}{2}(\beta^{-1} - \alpha^{-1})$  centered on the real axis of the complex plane at  $[-\frac{1}{2}(\alpha^{-1} + \beta^{-1}), 0]$  (see Fig. 3).

*Proof:* Note first that

$$\int_0^{\infty} |u(t)|^2 dt \le \max(\beta^2, |\alpha|^2) \int_0^{\infty} |f(t)|^2 dt,$$

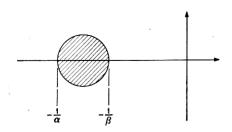


Fig. 2 — Location of the "critical circle"  $C_1$  in the complex plane  $(\alpha > 0)$ . The feedback system is  $\mathfrak{L}_2$ -stable if the locus of  $K(i\omega)$  for  $-\infty < \omega < \infty$  lies outside  $C_1$  and does not encircle  $C_1$ .

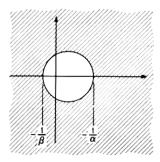


Fig. 3 — Location of the ''critical circle''  $C_2$  in the complex plane  $(\alpha < 0)$ . The feedback system is  $\mathfrak{L}_2$ -stable if the locus of  $K(i\omega)$  for  $-\infty < \omega < \infty$  is contained within  $C_2$ .

and hence, by a well-known result,

$$\int_{0}^{\infty} \left| \int_{0}^{t} k(t - \tau) u(\tau) d\tau \right|^{2} dt \leq \left( \int_{0}^{\infty} |k(t)| dt \right)^{2} \int_{0}^{\infty} |u(t)|^{2} dt$$

$$\leq \max \left( \beta^{2}, |\alpha|^{2} \right) \left( \int_{0}^{\infty} |k(t)| dt \right)^{2} \int_{0}^{\infty} |f(t)|^{2} dt.$$

Using Minkowski's inequality,

$$\left(\int_{0}^{\infty} |v(t)|^{2} dt\right)^{\frac{1}{2}} \leq \left(\int_{0}^{\infty} \left| \int_{0}^{t} k(t - \tau)u(\tau) d\tau \right|^{2} dt\right)^{\frac{1}{2}} \\
+ \left(\int_{0}^{\infty} |g_{2}(t)|^{2} dt\right)^{\frac{1}{2}} \leq \max(\beta, |\alpha|) \int_{0}^{\infty} |k(t)| dt \\
\cdot \left(\int_{0}^{\infty} |f(t)|^{2} dt\right)^{\frac{1}{2}} + \left(\int_{0}^{\infty} |g_{2}(t)|^{2} dt\right)^{\frac{1}{2}}.$$

Consider now the relation between  $(g_1 + g_2)$  and f:

$$g_1(t) + g_2(t) = f(t) + \int_0^t k(t - \tau) \psi[f(\tau), \tau] d\tau, \quad t \ge 0$$

and suppose that

$$\int_{0}^{\infty} |g_{1}(t) + g_{2}(t)|^{2} dt < \infty.$$

According to the results of Ref. 1, our assumptions\* imply that there

$$\int_0^t |f(\tau)|^2 d\tau < \infty$$

<sup>\*</sup> In Ref. 1 it is assumed that

exists a positive constant  $\rho_1$  (which does not depend upon  $g_1$  or  $g_2$ ) such that

$$\int_0^\infty |f(t)|^2 dt < \rho_1 \int_0^\infty |g_1(t) + g_2(t)|^2 dt$$

provided that, with

$$K(s) = \int_0^\infty k(t)e^{-st} dt$$

and  $\omega = \text{Im}[s]$ ,

(i)  $1 + \frac{1}{2}(\alpha + \beta)K(s) \neq 0$  for  $Re[s] \geq 0$ , and

 $(ii) \ \tfrac{1}{2}(\beta-\alpha) \max_{-\infty<\omega<\infty} |K(i\omega)[1+\tfrac{1}{2}(\alpha+\beta)K(i\omega)]^{-1}| < 1.$ 

Thus the feedback system of Fig. 1 is  $\mathcal{L}_2$ -stable if conditions (i) and (ii) are satisfied.

According to the well-known theorem of complex-function theory that leads to the Nyquist criterion, condition (i) is satisfied if (and only if) the polar plot of  $K(i\omega)$  for  $-\infty < \omega < \infty$  does not encircle or pass through the point  $[-2(\alpha + \beta)^{-1},0]$ . It can easily be verified that condition (ii) is met if one of the following three conditions is satisfied.

- (a)  $\alpha > 0$ , and the locus of  $K(i\omega)$  for  $-\infty < \omega < \infty$  lies outside the circle  $C_1$  of radius  $\frac{1}{2}(\alpha^{-1} \beta^{-1})$  centered in the complex plane at  $[-\frac{1}{2}(\alpha^{-1} + \beta^{-1}), 0]$ .
  - (b)  $\alpha = 0$ , and  $\text{Re}[K(i\omega)] > -\beta^{-1}$  for all real  $\omega$ .
- (c)  $\alpha < 0$ , and the locus of  $K(i\omega)$  for  $-\infty < \omega < \infty$  is contained within the circle  $C_2$  of radius  $\frac{1}{2}(\beta^{-1} \alpha^{-1})$  centered in the complex plane at  $[-\frac{1}{2}(\alpha^{-1} + \beta^{-1}), 0]$ .

If  $\alpha > 0$ , the point  $[-2(\alpha + \beta)^{-1},0]$  lies on the real-axis diameter of  $C_1$ , while if condition (b) or (c) is met, it is impossible for the polar plot of  $K(i\omega)$  to encircle the point  $[-2(\alpha + \beta)^{-1},0]$ . Therefore, the conditions of the theorem guarantee that the feedback system is  $\mathfrak{L}_2$ -stable.

## Remarks

With regard to the necessity of our sufficient conditions for  $\mathcal{L}_2$ -stability, consider, for example, the case in which  $\alpha > 0$  and suppose, for simplicity, that v and u are related by a differential equation of the type mentioned in Section II. Then, a moment's reflection shows that there exists a  $\psi(x,t)$ , in fact a  $\psi(x,t)$  which is independent of t and linear in x,

for all finite t > 0. Our assumption that (2) is satisfied for all finite t > 0 implies that this condition is met.

that satisfies our assumptions and for which the feedback system is not  $\mathfrak{L}_2$ -stable, provided that for some value of  $\omega$ ,  $K(i\omega)$  is a point on the real-axis diameter of  $C_1$ . This clearly shows that the condition is in the correct "ball park." Similar remarks can be made concerning our conditions for the cases in which  $\alpha < 0$  and  $\alpha = 0$ .

# IV. FURTHER PROPERTIES OF THE FEEDBACK SYSTEM OF FIG. 1

It is possible to say much more about the properties of the feedback system on the basis of frequency-domain information if our assumptions regarding  $\psi(x,t)$  are strengthened.

For example, suppose that

$$\alpha \le \frac{\psi(x_1,t) - \psi(x_2,t)}{x_1 - x_2} \le \beta, \qquad \psi(0,t) = 0$$
 (4)

for  $t \ge 0$  and all real  $x_1 \ne x_2$ , and that one of the three conditions of our theorem is met. Let  $g_1$  and  $\hat{g}_1$  denote two arbitrary input functions such that

$$\int_0^t |g_1(\tau)|^2 d\tau < \infty \text{ and } \int_0^t |\hat{g}_1(\tau)|^2 d\tau < \infty$$

for all finite t > 0, and

$$\int_0^\infty |g_1(\tau) - \hat{g}_1(\tau)|^2 d\tau < \infty.$$

Let v and  $\hat{v}$ , respectively, denote the (assumed well defined) responses due to  $g_1$  and  $\hat{g}_1$ . Then if

$$\int_0^t |v(\tau)|^2 d\tau < \infty \quad \text{and} \quad \int_0^t |\hat{v}(\tau)|^2 d\tau < \infty$$

for all finite t > 0, and the assumptions of Section II are met, it follows\* that

$$\int_0^\infty |v(\tau) - \hat{v}(\tau)|^2 d\tau < \infty$$

and that there exists a positive constant  $\lambda$  (which does not depend upon  $g_1$  or  $\hat{g}_1$ ) such that

$$\int_0^{\infty} |v(\tau) - \hat{v}(\tau)|^2 d\tau \le \lambda \int_0^{\infty} |g_1(\tau) - \hat{g}_1(\tau)|^2 d\tau.$$

<sup>\*</sup> Consider Theorem 1 of Ref. 6 with  $h_1(t) = f_1(t) = 0$  for t < 0.

Suppose now that  $\psi(x,t)$  satisfies (4) and is either independent of t or periodic in t with period T for each x, and that one of the three conditions of our theorem is met. Assume that the initial-condition function  $g_2(t)$  approaches zero as  $t \to \infty$ , and that the input  $g_1(t)$  applied at t = 0 is a bounded periodic function with period T. Then it can be shown\* that there exists a bounded periodic function p, with period T, which is independent of  $g_2$  and such that the (assumed well defined) response v(t) approaches p(t) as  $t \to \infty$ , provided that the conditions of Section II are met, (2) is satisfied for all finite t > 0, and

$$\int_0^\infty \left| \int_t^\infty |k(\tau)| d\tau \right|^2 dt < \infty, \qquad \int_0^\infty |(1+t)k(t)|^2 dt < \infty.$$
 (5)

Observe that the conditions of (5) are satisfied if u and v are related by a differential equation of the form described in Section II.

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<sup>\*</sup> See Theorem 3 of Ref. 6.